

# Result Analysis of Carry Skip Adder using High-Speed skips logic at different levels

Rajesh sahu <sup>1</sup>, Deepak kirar <sup>2</sup>

<sup>1</sup> Mtech. Scholar, ECE, VIST, Bhopal, rajsahu512@gmail.com, India;

<sup>2</sup> H.O.D, ECE, VIST, Bhopal, E-mail, India;

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**Abstract** – A carry skip adder (CSKA) structure has the high speed and very low power consumption. The speed of the structure is achieved by concatenation of all the blocks. The incrimination blocks are used to improve the efficiency of the carry skip adder structure. In existing method multiplexer logic is used, the proposed structure uses the AND-OR-Invert (AOI) and OR-AND-Invert (OAI) for the skip logic. The carry skip adder structure is realized with both fixed stage size and variable stage size where the delay is reduced, and speed is improved. A hybrid variable latency extension lowers the power consumption without affecting the speed of the circuit. The results are obtained using XILINX 14.3 and it gives improvements in the delay and energy of the structures. In addition to this structure, the power–delay product was low among all the structures, while having its energy–delay product was almost same as that of the conventional structure. Simulations on the proposed structure by using hybrid variable latency CSKA reduces the power consumption compared with the previous works and it produces a high speed.

**Keywords:** Carry skip adder (CSKA), energy efficient, high performance, hybrid variable latency Adders,

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## I. Introduction

One of the effective techniques to lower the power utilization of digital circuits is to reduce the supply voltage because of quadratic dependence of the switching energy on the voltage. Moreover, the sub threshold current, that is the main leakage element in OFF devices, has an exponential dependence on the supply voltage level through the drain-induced barrier lowering impact. Depending on the amount of the supply voltage reduction, the operation of ON devices may reside in the super threshold, near-threshold, or sub threshold regions. Working in the super threshold region provides us with lower delay and better switching and leakage powers compared with the near/sub threshold regions. In the sub threshold region, the gate delay and leakage power exhibit exponential dependences on the supply and threshold voltages. Moreover, these voltages are (potentially) subject to process and environmental variations in the nano-scale technologies. The variations increase uncertainties in the said performance parameters. Additionally, the small sub threshold current causes a large delay for the circuits operating in the sub threshold region.

Recently, the near-threshold region has been considered as a section that gives an additional desirable trade-off purpose between delay and power dissipation

Compared with that of the sub threshold one, as a result of it leads to lower delay compared with the sub threshold region and considerably lowers switching and leakage powers compared with the super threshold region. Additionally, near-threshold operation, that uses supply voltage, levels near the threshold voltage of transistors, suffers significantly less from the process and environmental variations compared with the sub threshold region.

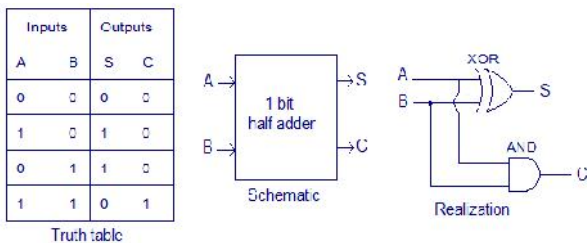
The dependence of the power (and performance) on the supply voltage has been the motivation for style of circuits with the feature of dynamic voltage and frequency scaling. In these circuits, to reduce the energy utilization, the system might change the voltage (and frequency) of the circuit based on the work requirement. For these systems, the circuit should be able to operate underneath a wide range of provide voltage levels. Of course, achieving higher speeds at lower provide voltages for the computational blocks, with the adder mutually the main components, could be crucial in the style of high-speed, yet energy efficient, processors. Adders are a key building block in arithmetic and logic units (ALUs) and hence increasing their speed and reducing their power/energy consumption strongly affect the speed and power utilization of processors. There are several works on the subject of optimizing the speed and power of these units that are reported in. Obviously, it's highly desirable to realize higher speeds at low-power/energy utilizations, which is a challenge for the designers of general purpose processors.

## II. Theory

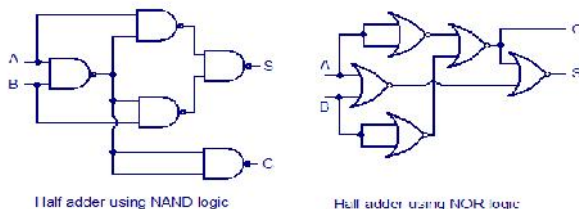
### II.1. The Half adder

To understand what is a half adder you need to know what is an adder first. Adder circuit is a combinational digital circuit that is used for adding two numbers. A typical adder circuit produces a sum bit (denoted by S) and a carry bit (denoted by C) as the output. Typically adders are realized for adding binary numbers but they can be also realized for adding other formats like BCD (binary coded decimal, XS-3 etc. Besides addition, adder circuits can be used for a lot of other applications in digital electronics like address decoding, table index calculation etc.

Half adder is a combinational arithmetic circuit that adds two numbers and produces a sum bit (S) and carry bit (C) as the output. If A and B are the input bits, then sum bit (S) is the X-OR of A and B and the carry bit (C) will be the AND of A and B. From this it is clear that a half adder circuit can be easily constructed using one X-OR gate and one AND gate. Half adder is the simplest of all adder circuit, but it has a major disadvantage. The half adder can add only two input bits (A and B) and has nothing to do with the carry if there is any in the input. So if the input to a half adder have a carry, then it will be neglected it and adds only the A and B bits. That means the binary addition process is not complete and that's why it is called a half adder. The truth table, schematic representation and XOR//AND realization of a half adder are shown in the figure below.



NAND gates or NOR gates can be used for realizing the half adder in universal logic and the relevant circuit diagrams are shown in the figure below.



### II.2. Full Adder.

This type of adder is a little more difficult to implement than a half-adder. The main difference between a half-adder and a full-adder is that the full-adder has three inputs and two outputs. The first two inputs are A and B and the third input is an input carry designated as CIN. When full adder logic is designed we will be able to string eight of them together to create a byte-wide adder and cascade the carry bit from one adder

to the next. The output carry is designated as COUT and the normal output is designated as S. The truth-table is given below.

INPUTS		OUTPUTS		
A	B	CIN	COUT	S
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

From the above truth-table, the full adder logic can be implemented. We can see that the output S is an EXOR between the input A and the half-adder SUM output with B and CIN inputs. We must also note that the COUT will only be true if any of the two inputs out of the three are HIGH.

Thus, we can implement a full adder circuit with the help of two half adder circuits. The first will half adder will be used to add A and B to produce a partial Sum. The second half adder logic can be used to add CIN to the Sum produced by the first half adder to get the final S output. If any of the half adder logic produces a carry, there will be an output carry. Thus, COUT will be an OR function of the half-adder Carry outputs.

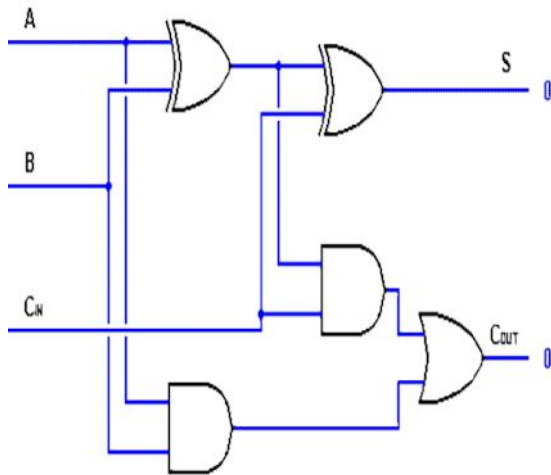


Figure: Full Adder Circuit

### II.3. Ripple carry adder

A ripple carry adder is a digital circuit that produces the arithmetic sum of two binary numbers. It can be constructed with full adders connected in cascaded, with the carry output from each full adder connected to the carry input of the next full adder in the chain. Figure 3 shows the interconnection of four full adder (FA) circuits to provide a 4-bit ripple carry adder. Notice from Figure 3 that the input is from the right side because the first cell traditionally represents the least significant bit (LSB). Bits  $a_0$  and  $b_0$  in the figure represent the least significant bits of the numbers to be added. The sum output is represented by the bits  $s_0$ - $s_3$ .

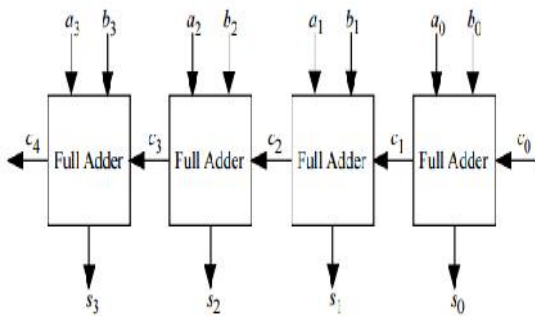


Figure: 4-bit full adder.

### II.4. Carry lookahead adder (CLA)

The carry look ahead adder (CLA) solves the carry delay problem by calculating the carry signals in advance, based on the input signals. It is based on the fact that a carry signal will be generated in two cases: (1) when both bits  $a_i$  and  $b_i$  are 1, or (2) when one of the two bits is 1 and the carry-in is 1. Thus, one can write,

$$c_{i+1} = a_i \cdot b_i + (a_i \oplus b_i) \cdot c_i$$

$$s_i = (a_i \oplus b_i) \oplus c_i$$

The above two equations can be written in terms of two new signals  $p_i$  and  $g_i$ , which are shown in Figure 4:

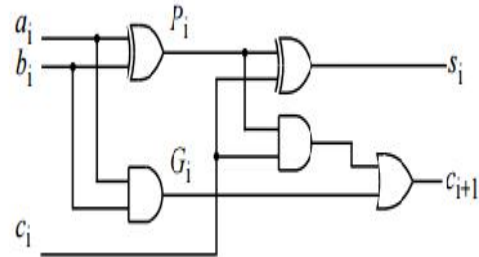


Figure: Full adder at stage  $i$  with  $p_i$  and  $g_i$  shown.

## III. Method

### III.1. Proposed CSKA structure

#### III.1.1. General description of the proposed structure

The structure is based on combining the concatenation and therefore the incrimination schemes with the Conv-CSKA structure, and hence, is denoted by CI-CSKA. It provides us with the ability to use simpler carry skip logics. The logic replaces 2:1 multiplexers by AOI/OAI compound gates (Fig. 2). The gates that consist of fewer transistors have lower delay, area, and smaller power consumption compared with those of the 2:1 multiplexer. Note that, in this structure, because the carry propagates through the skip logics, it becomes complemented. Therefore, at the output of the skip logic of even stages, the complement of the carry is generated. The structure contains a considerable lower propagation delay with a slightly smaller area compared with those of the conventional one. Note that whereas the power consumptions of the AOI (or OAI) gate are smaller than that of the multiplexer, the power consumption of the proposed CI-CSKA is a little more than that of the traditional one. This is often due to the increase in the number of the gates that imposes a higher wiring capacitance (in the noncritical paths).

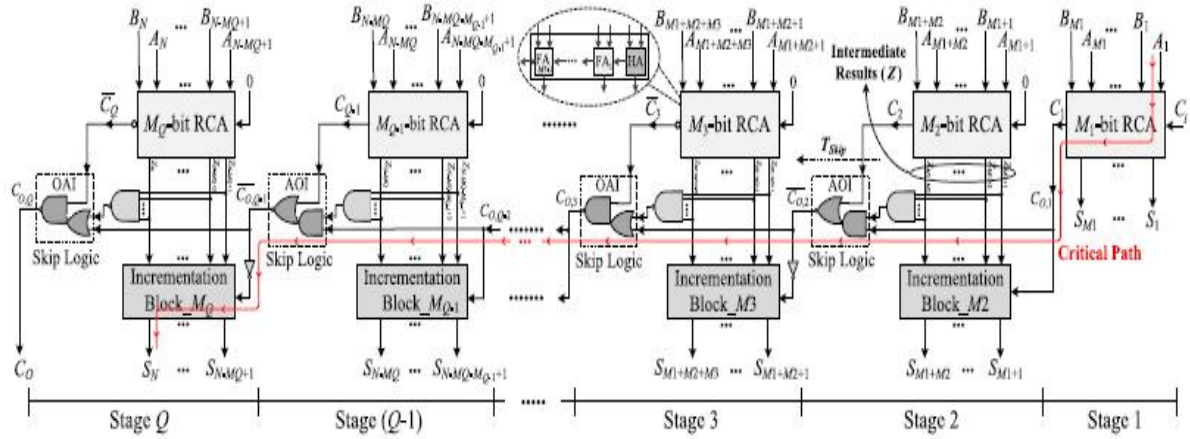
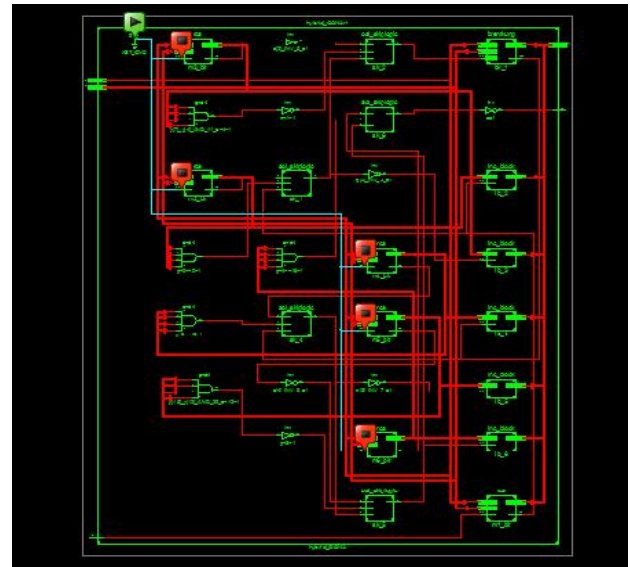
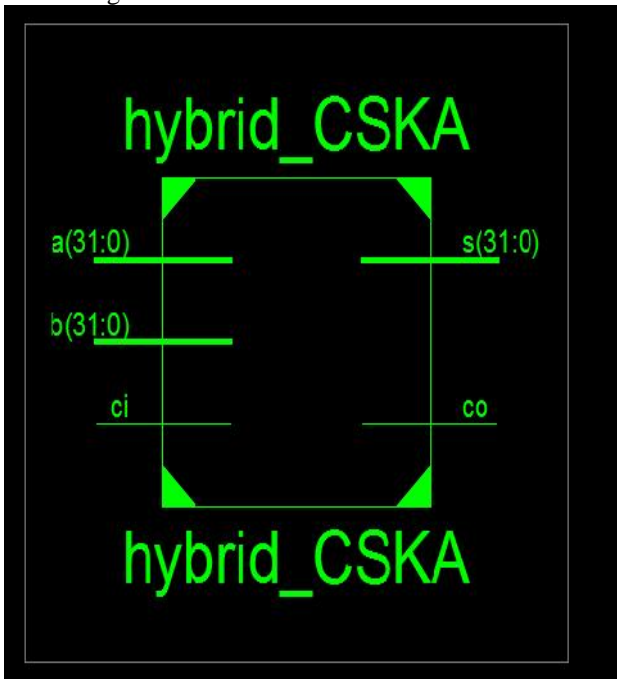


Fig.2. Proposed CI-CSKA structure

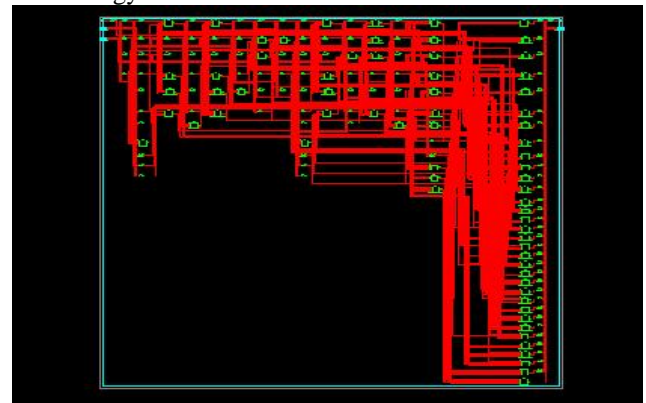
RTL schematic

#### IV. Result

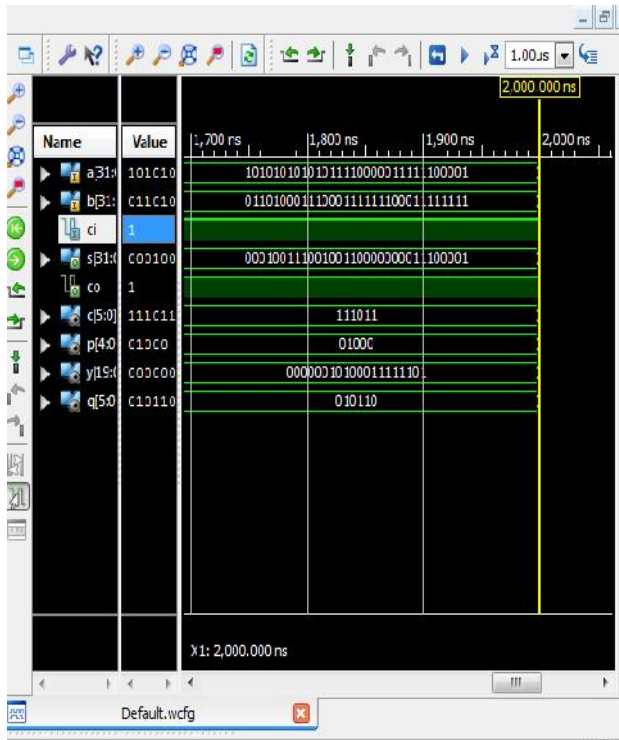
Block diagram



Technology schematic



## Simulation Results



## V. CONCLUSION

In this paper, a static CMOS CSKA structure known as CI-CSKA was proposed, that exhibits a higher speed and lower energy consumption compared with those of the conventional one. The speed enhancement was achieved by modifying the structure through the concatenation and incrimination techniques. Additionally, AOI and OAI compound gates were exploited for the carry skip logics. The efficiency of the proposed structure for both FSS and VSS was studied by comparison its power and delay with those of the Conv-CSKA, RCA, CIA, SQRT-CSLA, and KSA structures. The results revealed significantly lower PDP for the VSS implementation of the CI-CSKA structure over a wide range of voltage from super-threshold to near threshold. The results also suggested the CI-CSKA structure as a really good adder for the applications wherever both the speed and energy consumption are crucial. Additionally, a hybrid variable latency extension of the structure was planned. It exploited a modified parallel adder structure at the middle stage for increasing the slack time, which provided us with the chance for lowering the energy consumption by reducing the supply voltage. The effectiveness of this structure was compared versus those of the variable latency RCA, C2SLA, and hybrid C2SLA structures. Again, the suggested structure showed the lowest delay and PDP creating itself as a better candidate for high-speed low-energy applications.

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